

Elettra Sincrotrone Trieste



Noise contributions in eBPM systems

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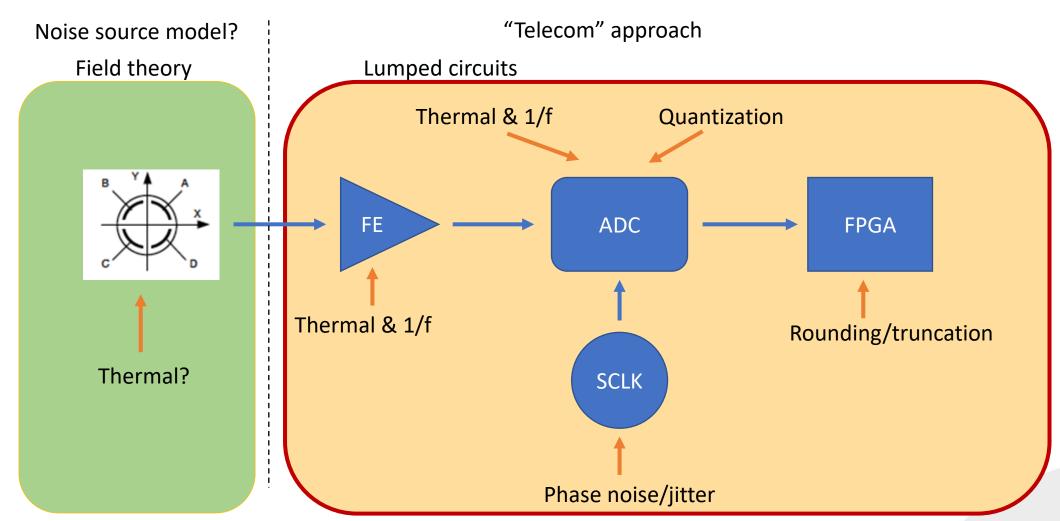
Noise in passive/active devices

- We refer as noise as a random process that can be described only through <u>statistical</u> <u>properties</u>, such as:
 - distribution (mean, variance, probability density function)
 - autocorrelation function and its Fourier transform (Power spectral density, PSD)
- Noise power can be estimated integrating the PSD over the given bandwidth

| Johnson – Nyquist (thermal) | Shot | Flicker or 1/f |
|--|---|--|
| Generated by thermal agitation of charge carriers in a conductor (resistors, channel in a FET, etc) Usually modelled as a gaussian process No correlation between samples PSD is flat (white noise), value is 4kTR for voltage noise | barriers (e.g. p-n junction) Usually modelled as a Poisson process No correlation between samples | Generated by impurities or generation-ricombination of charges in electronic devices Different processes: gaussian or chi-distributed Correlation between samples PSD is NOT flat, value is inversely proportional to frequency |



Noise in BPM acquisition chain

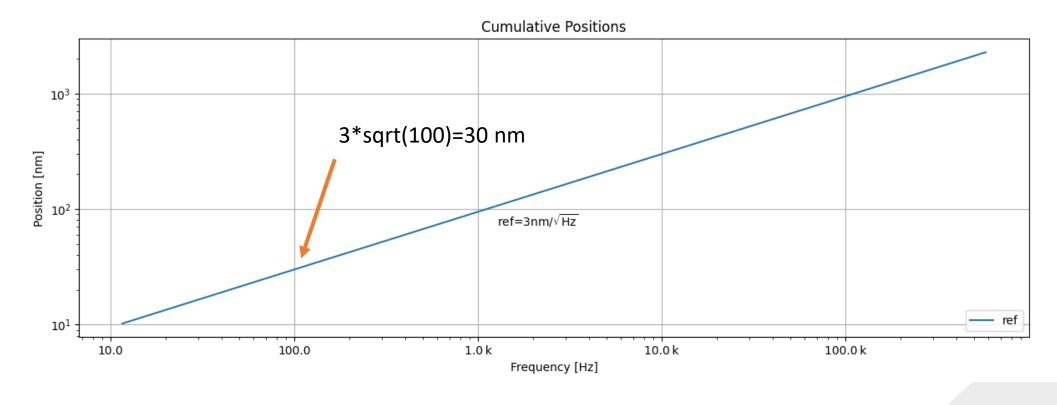






Noise evaluation in BPMs

- Easier to refer to noise on calculated positions
- graphs: cumulative sum of PSD vs frequency







Noise evaluation in BPMs

- Test setup: RF gen + splitter (emulate a stable beam)
- Significant 1/f noise on our BPMs
- Pilot tone compensation can cancel it out
- Coming from ADCs?
 - No difference w & w/o front end -> PTFE has white noise
 - No difference at different frequencies -> not related to sampling clock jitter
 - Could be a ADC "signature"? -> different between various ADC models? (tests in progress)

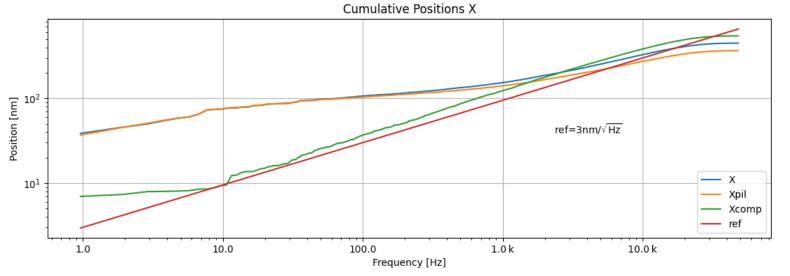




Cumulative Spectral Density

1.0 k

10.0 k





Frequency [Hz]

100.0

100 kS/s data rate

Red trace: white noise reference of 3nm/sqrt(Hz)

Blue trace: carrier Orange trace: pilot

Green trace: compensated

position

Noise of non-compensated position higher than expected

Compensated position follows theoretical prediction

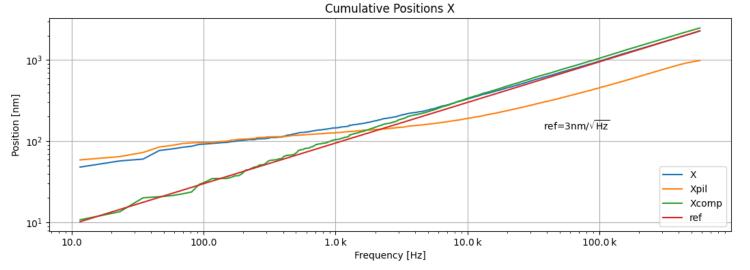


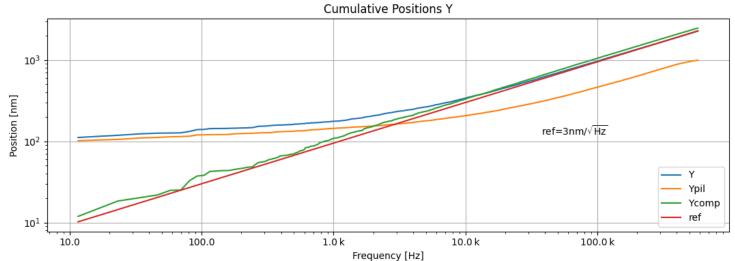
1.0

10.0



Cumulative Spectral Density





1.156 MS/s data rate - TbT

Noise PSD is flat over 7/10 kHz

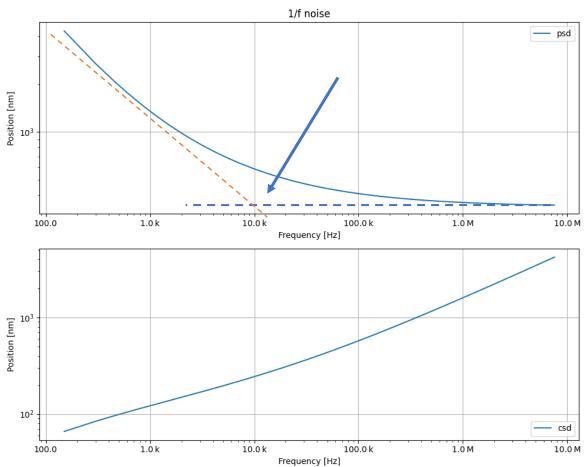
Expanding bandwidth helps to understand the phenomena

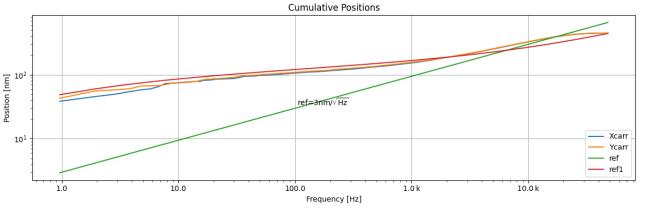
Noise on pilot positions is lower due to enhanced filtering (reduced BW)





Sincrotrone Trieste 1/f fitting





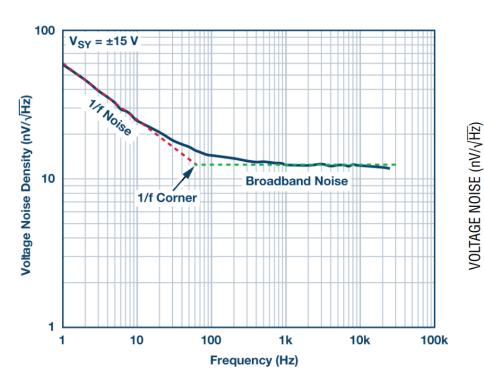
Fit 1/f noise (red trace):

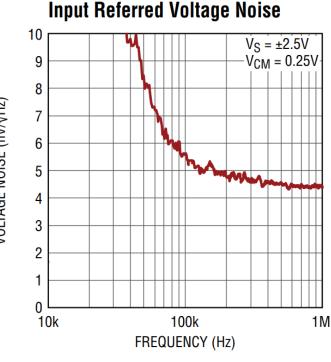
PSD:
$$S(f) = \frac{k^2 F_c}{f} + k^2$$
, k = 3nm/sqrt(Hz)

Corner frequency: about 10 kHz



Examples of 1/f noise in opamps





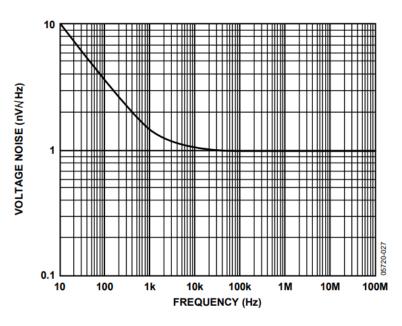


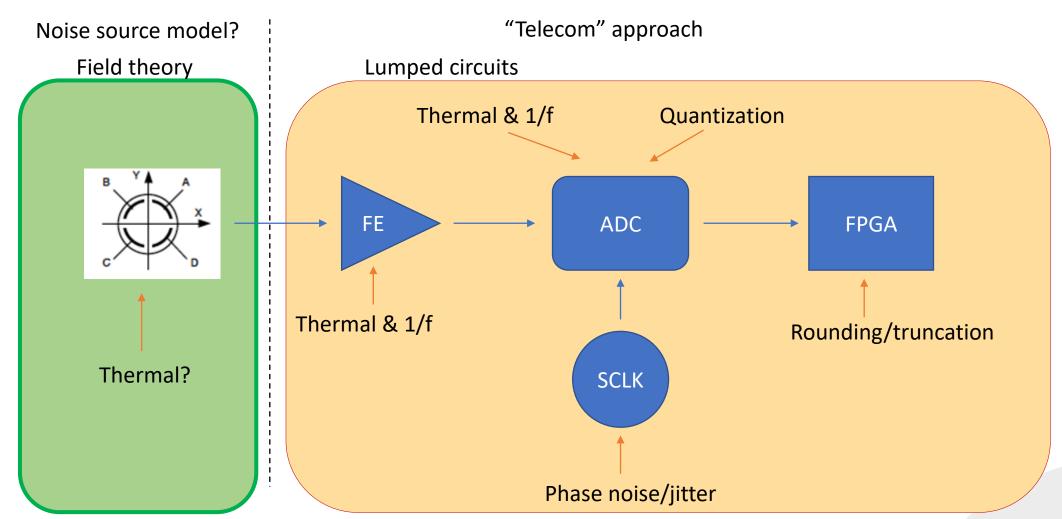
Figure 13. Voltage Noise vs. Frequency



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Noise in BPM acquisition chain







S/N and position resolution

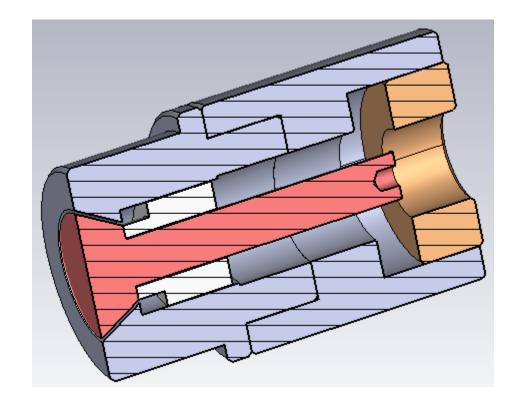
- Measurements are limited by (thermal) noise
- Signal power: TEM mode propagation implies that voltage and current concepts for lumped circuits can be defined at the signal port of the Pickup
- Noise power: what kind of electric source model matches the Pickup?
- Several articles/papers assume a thermal noise generator (Thevenin equivalent circuit)
- Thermal movement of electrons, independently from each other, in conducting material ($\sigma > 0$)





Button type pickup

 Button type PU: what is R? How a coaxial propagation path can be modeled as a noise source?

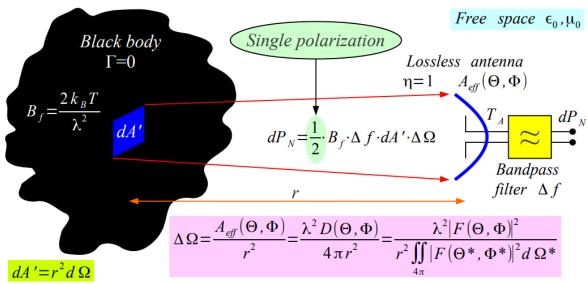






Sincrotrone Trieste Antenna theory

 Some references (M. Vidmar, Noise in Radio/Optical Communications, IBIC 2108; thanks to B. Roche for pointing it out)



$$P_{N} = \iint_{A'} \frac{1}{2} \cdot B_{f} \cdot \Delta f \cdot dA' \cdot \Delta \Omega = \iint_{4\pi} \frac{1}{2} \cdot \frac{2k_{B}T(\Theta, \Phi)}{\lambda^{2}} \cdot \Delta f \cdot r^{2} d\Omega \cdot \frac{\lambda^{2} |F(\Theta, \Phi)|^{2}}{r^{2} \iint_{A\pi} |F(\Theta^{*}, \Phi^{*})|^{2} d\Omega^{*}}$$

$$P_{N} = \Delta f k_{B} \frac{\iint_{4\pi} T(\Theta, \Phi) |F(\Theta, \Phi)|^{2} d\Omega}{\iint_{4\pi} |F(\Theta, \Phi)|^{2} d\Omega} = \Delta f k_{B} T_{A}$$

$$T \iint_{4\pi} |F(\Theta, \Phi)| |F(\Theta, \Phi)|^{2} d\Omega$$

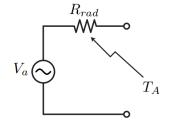
$$T_{A} = \frac{\iint_{4\pi} T(\Theta, \Phi) |F(\Theta, \Phi)|^{2} d\Omega}{\iint_{4\pi} |F(\Theta, \Phi)|^{2} d\Omega}$$

 $\frac{1}{4\pi} = \frac{1}{4\pi}$ 4 - Received thermal-noise power

Power radiated by a physical object: $P_N = kT_NB$

Power collected by a **lossless** antenna: $P_A = SA_eB = kT_AB$, depends on solid angles of source and receiver (radiation pattern) - T_A is **not related** to physical temperature of the antenna

Equivalent noise circuit: we shall consider the radiation resistance of the antenna (ratio between E and H):



$$V_a^2(f) = 4kT_A R_{rad} B$$

With matched load: $P_L = kT_AB = P_A$

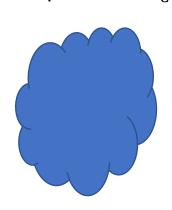


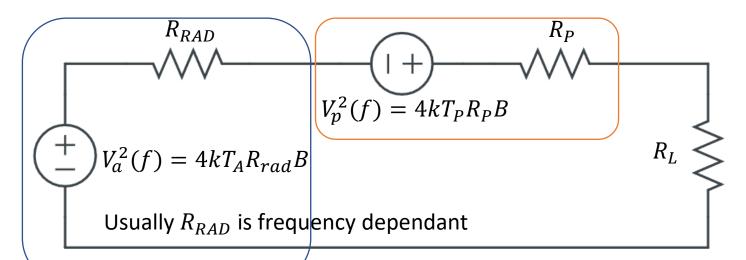
Pickup – proposed equivalent circuit

Vacuum chamber (is it radiating noise power? At which temperature? T_C)

Pickup (antenna with a narrow directivity, T_A and a given R_{RAD})

Pickup losses from conductor and dielectric? T_P ?





Load impedance – usually very different from R_{RAD} (coax cable omitted)

Voltage noise on
$$R_L$$
: $V_L^2 = V_a^2 \left(\frac{R_L}{R_L + R_{RAD} + R_P}\right)^2 + V_p^2 \left(\frac{R_L}{R_L + R_{RAD} + R_P}\right)^2$

Depends on:

- Temperatures (real and equivalent)
- Ratio between impedances



Thermal noise standards (NIST)

- Johnson noise thermometry based on available power: T=Pa/kB
- The power density is assumed constant in the measurment bandwidth
- For the evaluation of the noise along the BPM acquisition chain a coaxial thermal noise generator could be used as noise source.





Sincrotrone Trieste Questions

- Have you ever observed 1/f noise on calculated positions from BPMs? Could you define a clear "electronic" source?
- How can be modelled the noise from the pickups? Or can they be considered "noiseless" at the end?
 - References in "traditional" articles/papers take for granted the thermal noise behaviour with matched impedances (available power P=kTB)





Thank you!





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